

UNISONIC TECHNOLOGIES CO., LTD

UA7527

Preliminary

LINEAR INTEGRATED CIRCUIT

POWER FACTOR CONTROLLER

DESCRIPTION

The UTC UA7527 is a simple and high performance active power factor corrective controller for boost PFC application which operates in the critical conduction mode. The UTC UA7527 is optimized for electronic ballasts, low power and high density power supplies which require minimum board area reduced component count and low power dissipation. Internal R/C filter eliminates the need for an external R/C filter. Special circuitry has also been added to prevent no load runaway conditions. The output drive clamping circuit limits the overshoot of the power MOSFET gate drive despite the supply voltage. This greatly enhances the system reliability.

FEATURES

- * Internal Start-up Timer
- * Very Precise Adjustable Output Over Voltage Protection
- * Zero Current Detector
- * Quadrant Multiplier
- * Internal R/C Filter Eliminates the Need for an External R/C Filter
- * Trimmed 1.5% Internal Band Gap Reference
- * Under Voltage Lockout with 3V of Hysteresis
- * Totem Pole Output With High State Clamp
- * Low Start-up and Operating Current

ORDERING INFORMATION

| Ordering Number | Package | Packing |
|-----------------|---------|-----------|
| UA7527G-S08-R | SOP-8 | Tape Reel |

| UA7527G-S08-R (1)Packing Type (2)Package Type | (1) R: Tape Reel (2) S08: SOP-8 |
|---|------------------------------------|
| (3)Green Package | (3) G: Halogen Free and Lead Free |

MARKING





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PIN CONFIGURATION



PIN DESCRIPTION

| PIN NO. | PIN NAME | DESCRIPTION |
|---------|------------------|---|
| 1 | INV | The inverting input of the error amplifier. The output of the boost converter should be resistively divided to 2.5V and connected to this pin. |
| 2 | EA_OUT | The error amplifier output. A feedback compensation network is placed between this pin and the INV pin. |
| 3 | MULT | Input to the multiplier stage. Full-wave rectified AC voltage is divided into a voltage less than 2V and is connected to this pin. |
| 4 | CS | Input of the Pulse-Width Modulation comparator. Current is sensed in the boost stage MOSFET by a resistor in the source lead. An internal R / C filter can filter out any high frequency noise. |
| 5 | I _{DET} | Input to zero current detection. |
| 6 | GND | The IC ground. |
| 7 | OUT | Gate driver output. This pin provides an output to an external Power MOSFET with peak current of 500mA. |
| 8 | V _{cc} | The positive supply of the device. |

UNISONIC TECHNOLOGIES CO., LTD

BLOCK DIAGRAM





Preliminary LINE

LINEAR INTEGRATED CIRCUIT

■ ABSOLUTE MAXIMUM RATING (T_A=25°C, unless otherwise specified)

| PARAMETER | SYMBOL | RATINGS | UNIT |
|--|-----------------------------------|-----------|------|
| Supply Voltage | V _{CC} | 30 | V |
| Multiplier, Error Amp and Comparator Input Voltages | V _{IN} | -0.3 ~ 6 | V |
| Peak Drive Output Current | I _{OH} , I _{OL} | ±500 | mA |
| Output Clamping Diodes Current $V_0 > V_{CC}$ or $V_0 < -0.3V$ | I _{CLAMP} | ±10 | mA |
| Detect Clamping Diodes Current | IDET | ±10 | mA |
| Power Dissipation | PD | 0.8 | W |
| Junction Temperature | T_J | 150 | °C |
| Operating Temperature | T _{OPR} | -25 ~ 125 | °C |
| Storage Temperature | T _{STG} | -65 ~ 150 | °C |

Note: Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

■ **TEMPERATURE CHARACTERISTICS** (-25°C ≤ T_A ≤ 125°C)

| PARAMETER | SYMBOL | RATINGS | UNIT |
|--|------------------|---------|------|
| Temperature Stability for Reference Voltage (VREF) | ΔV_{REF} | 20 | mV |
| Temperature Stability for Multiplier Gain (K) | ΔΚ/ΔΤ | -0.2 | %/°C |

■ ELECTRICAL CHARACTERISTICS (V_{CC} = 14V, -25°C ≤ T_A ≤ 125°C, unless otherwise specified.)

| PARAMETER | SYMBOL | TEST CONDITIONS | MIN | TYP | MAX | UNIT | |
|--|----------------------|---|------------------|------|---------------------------|------|--|
| UNDER VOLTAGE LOCKOUT SECTION | | | | | | | |
| Start Threshold Voltage | V _{TH(ST)} | V _{CC} Increasing | 10.5 | 11.5 | 12.5 | V | |
| UV lockout Hysteresis | HY _(ST) | | 2 | 3 | 4 | V | |
| SUPPLY CURRENT SECTION | | | | | | | |
| Start-up Supply Current | I _{ST} | $V_{CC} = V_{TH(ST)} - 0.2V$ | 10 | 60 | 100 | uA | |
| Operating Supply Current | Icc | Output not switching | | 3 | 6 | mA | |
| Operating Current at OVP | I _{CC(OVP)} | V _{INV} = 3V | | 1.7 | 4 | mA | |
| Dynamic Operating Supply Current | I _{DCC} | 50kHz, CI = 1nF | | 4 | 8 | mA | |
| ERROR AMPLIFIER SECTION | | | | | | | |
| | V | I _{REF} = 0mA, T _A = 25°C | 2.465 | 2.5 | 2.535 | V | |
| Voltage Feedback linput Threshold | V REF | -25 ≤ T _A ≤ 125°C | 2.44 | 2.5 | 2.56 | V | |
| Line Regulation | ΔV_{REF1} | $14V \le V_{CC} \le 25V$ | | 0.1 | 10 | mV | |
| Temperature Stability of V _{REF} (Note 1) | ΔV_{REF3} | -25 ≤ T _A ≤ 125°C | | 20 | | mV | |
| Input Bias Current | I _{B(EA)} | | -0.5 | | 0.5 | uA | |
| | ISOURCE | $V_{M2} = 4V$ | -2 | -4 | | mA | |
| Output Current | I _{SINK} | $V_{M2} = 4V$ | 2 | 4 | | mA | |
| Output Upper Clamp Voltage (Note 2) | V _{EAO(H)} | I _{SOURCE} = 0.1mA | | 6 | | V | |
| Output Lower Clamp Voltage (Note 3) | V _{EAO(L)} | I _{SINK} = 0.1mA | | 2.25 | | V | |
| Large Signal Open Loop gain (Note 4) | Gv | | 60 | 80 | | dB | |
| Power Supply Rejection Ratio (Note 5) | PSRR | $14V \le V_{CC} \le 25V$ | 60 | 80 | | dB | |
| Unity Gain Bandwidth (Note 6) | GBW | | | 1 | | MHz | |
| Slew Rate (Note7) | SR | | | 0.6 | | V/us | |
| MULTIPLIER SECTION | | | ~ | | | | |
| Input Bias Current (Pin3) | I _{B (M)} | 1 | -0.5 | | 0.5 | uA | |
| M1 Input Voltage Range (Pin3) | ΔV_{M1} | TR2 V | 0 | | 3.8 | V | |
| M2 Input Voltage Range (Pin2) | ΔV_{M2} | 唐 有 688 | V _{REF} | | V _{REF} +2. 5 | V | |
| Multiplier Gain (Note8) | К | V _{M1} = 1V, V _{M2} = 3.5V | 0.36 | 0.44 | 0.52 | 1/V | |
| Maximum Multiplier Output Voltage | V _{OMAX(M)} | $V_{INV} = 0V, V_{M1} = 4V$ | 1.65 | 1.8 | 1.95 | V | |
| Multiplier Gain Stability (Note 9) | ΔΚ/ΔΤ | -25 ≤ T _A ≤ 125°C | | -0.2 | | %/°C | |
| | CN | N | | | | | |

ELECTRICAL CHARACTERISTICS(Cont.)

| PARAMETER | | SYMBOL | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|------------------------------------|------|----------------------|---|------|------|-----|------|
| CURRENT SENSE SECTION | | | | | | | |
| Input Offset Voltage (Note 8) | | V _{IO(CS)} | V _{M1} =0V, V _{M2} = 2.2V | -10 | 3 | 10 | mV |
| Input Bias Current | | I _{B(CS)} | $0V \le V_{CS} \le 1.7V$ | -1 | -0.1 | 1 | uA |
| Current Sense Delay to Output | | t _{D(CS)} | (Note 11) | | 200 | 500 | ns |
| ZERO CURRENT DETECT SECTIO | N | _ | | | | | |
| Input Voltage Threshold | | V _{TH(DET)} | V _{DET} Increasing | 1.7 | 2 | 2.3 | V |
| Detect Hysteresis | | HY _(DET) | | 0.2 | 0.5 | 0.8 | V |
| Input Low Clamp Voltage | | V _{CLAMP} | I _{DET} = -100uA | 0.45 | 0.75 | 1 | V |
| Input High Clamp Voltage | | V _{CLAMP}) | I _{DET} = 3mA | 6.5 | 7.2 | 7.9 | V |
| Input Bias Current | | I _{B(DET)} | $1V \le V_{DET} \le 5V$ | -1 | -0.1 | 1 | uA |
| Input High/Low Clamp Diode Current | | I CLAMP(D) | (Note 12) | | | ±3 | mA |
| OUTPUT SECTION | | | | | | | |
| | High | V _{OH} | I _O = -10mA | 10.5 | 11 | | V |
| | Low | V _{OL} | I _O = 10mA | | 0.8 | 1 | V |
| Rising Time (Note 13) | | t _R | C _I = 1nF | | 130 | 200 | ns |
| Falling Time (Note 14) | | T _F | C _I = 1nF | | 50 | 120 | ns |
| Maximum Output Voltage | | V _{OMAX(O)} | V _{CC} = 20V, I _O = 100uA | 12 | 14 | 16 | V |
| Output Voltage with UVLO Activated | | V _{OMIN(O)} | V _{CC} = 5V , I _O = 100uA | | | 1 | V |
| RESTART TIMER SECTION | | | | | | | |
| Restart Time Delay | | t _{D(RST)} | V _{M1} = 1V, V _{M2} = 3.5V | | 150 | | us |
| OVER VOLTAGE PROTECTION SECTION | | | | | | | |
| Soft OVP Detecting Current | | I _{SOVP} | | 25 | 30 | 35 | uA |
| Dynamic OVP Detecting Current | | I _{DOVP} | | 35 | 40 | 45 | uA |
| Static OVP Threshold Voltage | | V _{OVP} | V _{INV} = 2.7V | 2.1 | 2.25 | 2.4 | V |

1~14.These parameters, although guaranteed, are not 100% tested in production. Multiplier Gain: $K = \frac{Pin4_Threshold}{V_{M1} \times (V_{M2} - V_{REF})} \dots (V_{M1} = Vpin3, V_{M2} = Vpin2)$ Note:



LINEAR INTEGRATED CIRCUIT

TYPICAL APPLICATION CIRCUIT



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