**UCL5811 Preliminary CMOS IC** 

# **PSR & PFC LED DRIVER** CONTROLLER

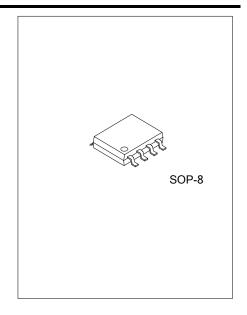
### **DESCRIPTION**

The UTC UCL5811 is a single-power stage, isolated and primary side offline LED lighting regulator which achieves high power factor.

The proprietary real-current control method can control the LED current accurately from the primary side information. It can significantly simplify the LED lighting system design by eliminating the feedback components and the opto-coupler.

The UTC UCL5811 integrates active power factor correction and works in Quasi Resonance mode (QRM) in order to reduce the MOSFET switching losses. With a building in 700V start-up MOSFET, IC can eliminate the power loss caused by start-up resistors to provide a high efficiency solution for lighting applications. The external programmable line voltage compensation provides a more precise output current throughout the universal AC input voltage range. The leading edge blanking circuit on the CS/FB input removes the signal glitch and results in reduced external components and system cost. The multi-protection features of UTC UCL5811 greatly enhance the system reliability and safety.

The UTC UCL5811 V<sub>DD</sub> and output over voltage protection; output short circuit protection, cycle-by-cycle current limit and secondary peak current protection on CS pin, VDD UVLO andauto-restart and over-temperature protection. The driver output voltage is clamped at 18V to protect the external power MOSFET.



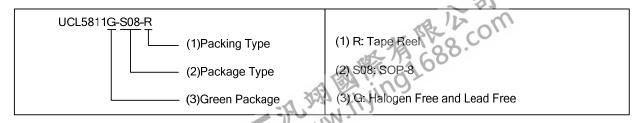
### **FEATURES**

- \* Single stage PFC
- \* Primary side regulation without Secondary Feedback
- \* Quasi Resonance (QR) mode with Fly-back topology
- \* Real-Current control to meet accurate output current
- \* Very less components
- \* Programmable input AC compensation
- \* Leading Edge Blanking on CS/FB pin

- \* Building in hysteresis OTP
- \* V<sub>DD</sub> over voltage protection
- \* Cycle by cycle current limiting on CS pin
- \* Secondary peak current protection on CS pin
- \* Output short to GND protection
- \* Output programmable over voltage protection
- \* FB and CS pins default protection

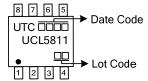
### **ORDERING INFORMATION**

Ordering Number	Package	Packing
UCL5811G-S08-R	SOP-8	Tape Reel

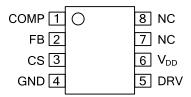


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# ■ MARKING



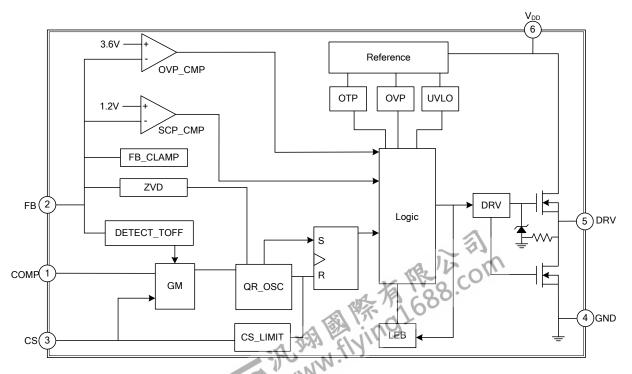
### **■ PIN CONFIGURATION**



### **■ PIN DESCRIPTION**

PIN NO.	PIN NAME	DESCRIPTION
1	COMP	Loop compensation pin. Connect a RC network across this pin and ground to stabilize the control loop.
2	FB	Detect output diode zero current to regulate output current. Connected to a resistor divider for sensing the reflected voltage from auxiliary winding.
3	CS	Current sense pin, a resistor connects to sense the MOSFET current.
4	GND	Power Ground.
5	DRV	Output to drive the external power MOSFET, Maximum Voltage is internally clamped to 18V.
6	$V_{DD}$	Power Supply pin.
7,8	NC	No connect.

# ■ BLOCK DIAGRAM



### **ABSOLUTE MAXIMUM RATING**

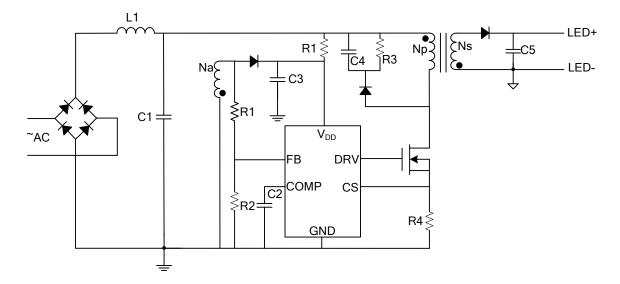
PARAMETER	SYMBOL	RATINGS	UNIT
Maximum supply voltage on V <sub>DD</sub> pin	$V_{DD}$	28	V
Input Voltage to FB Pin	$V_{FB}$	-0.3 ~ 6	V
Input Voltage to CS Pin	V <sub>CS</sub>	-0.3 ~ 6	V
Input Voltage to CMP Pin	$V_{CMP}$	-0.3 ~ 6	V
Maximum Operating Junction Temperature	TJ	+125	°C
Storage Temperature	T <sub>STG</sub>	-55 ~ <b>+</b> 150	°C

Note: Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

# **ELECTRICAL CHARACTERISTICS** (T<sub>C</sub>=25°C and V<sub>DD</sub>=15V unless otherwise specified.)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
V <sub>DD</sub> Pin			ı		I			
Operating Supply Current	I <sub>OP</sub>	Gate open		1		mA		
On Threshold Voltage	V <sub>DD_ON</sub>			16				
Off Threshold Voltage	V <sub>DD_OFF</sub>			9		V		
Logic Reset V <sub>DD</sub> Threshold Voltage	V <sub>DD_RST</sub>			6		V		
V <sub>DD</sub> Over Voltage Protection Level	V <sub>DD_OVP</sub>			25		V		
FB Pin	, <u> </u>	1	I		I			
FB Voltage When Sink Current	V <sub>FB_SINK</sub>	2mA sink current		5		V		
FB Voltage When Source Current	V <sub>FB_SOURCE</sub>	4mA source current	-0.1	GND	+0.1	V		
FB Zero Voltage Detect	V <sub>FB_ZVD</sub>			0.2		V		
FB Voltage When Output OVP	V <sub>FB_OVP</sub>			3.6		V		
FB Voltage When Trigger SCP	V <sub>FB_SCP</sub>			1		V		
CS Pin	1 <u>5_</u> 001		ı	1	I			
Cycle by Cycle Current Limited on CS	V <sub>CS</sub>	FB=0V		1		V		
Inner CS Voltage Vary with FB Source								
Current	$\Delta V_{CS}/\Delta I_{FB}$			TBD		mV/mA		
Leading-Edge Blanking Time	T <sub>LEB</sub>		300	400	500	ns		
COMP Pin				•				
Reference Voltage for OTA Input	$V_{REF}$			0.2		V		
COMP Maximal Sink Current	I <sub>COMP_SINK</sub>			50		uA		
COMP Maximal Source Current	I <sub>COMP_SOURCE</sub>			10		uA		
COMP Maximal Voltage	V <sub>COMP_MAX</sub>			4.0		V		
Oscillator				•				
Maximal Frequency	F <sub>OSC_MAX</sub>			130		kHz		
Minimal Frequency	Fosc_min			30		kHz		
DRV pin								
Rise Time	T <sub>Rise</sub>	C <sub>L</sub> =1nF		200		ns		
Fall Time	T <sub>Fall</sub>	C <sub>L</sub> =1nF		100		ns		
Over Temperature Drive Protection				_				
Over Temperature Protection	T <sub>OTP</sub>			145		°C		
OTP Hysteresis	T <sub>OTP_HYS</sub>	/		20		°C		
Over Temperature Protection Totp 145 °C OTP Hysteresis Totp_Hys 20 °C  UNISONIC TECHNOLOGIES CO., LTD 3 of 5 OW-R125-060.a								
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# **TYPICAL APPLICATION CIRCUIT**





### **■ FUNCTIONAL DESCRIPTION**

#### **Operating Description**

The UTC **UCL5811** is a primary side control offline LED controller that incorporates all the features for high performance LED lighting. LED current can be accurately controlled with the real current control method form the primary side information. Active Power Factor Correction (PFC) is included to eliminate the unwanted harmonic noise injected onto the AC line.

### Startup

During start-up, the current can charge up the  $V_{DD}$  hold capacitor. the turn-on and turn on and turn-off thresholds of UTC **UCL5811** are approximately 15V and 9V respectively. The 6V hysteresis voltage is implemented to prevent shutdown from a voltage dip during start-up.

### Quasi Resonance mode (QRM)

During the external power MOSFET on time  $(T_{ON})$  the rectified input voltage is applied across the primary side inductor  $(L_{M})$  and the primary current increases linearly from zero to the peak value  $(I_{PK})$ . When the external power MOSFET turns off, the energy stored in the inductor forces the secondary side diode to be turn-on, and the current of the inductor begins to decrease linearly from the peak value to zero. When the current decreases to zero, the parasitic resonant of inductor and all the parasitic capacitance makes the power MOSFET drain-source voltage decrease, this decreasing is also reflected on the auxiliary winding. The zero-current detector in FB pin generates the turn on signal of the external MOSFET when the FB voltage is lower than 0.2V and ensures the MOSFET turn on at a valley voltage.

As a result, there are virtually no primary switch turn-on losses and no secondary diode reverse-recover losses. It ensures high efficiency and low EMI noise.

### **Active Power Factor Correction (APFC)**

UTC **UCL5811** is designed with quasi-resonance and constant on time  $(T_{ON})$  to achieve high power factor under normal operation. The on time of UTC **UCL5811** vary with input AC voltage  $V_P \sin \omega t$  and load condition and its value is constant basically because of very large loop compensation capacitance on COMP pin. According to following equations.

$$I_{L-peak} = \frac{Vpsin\varpi t}{L_m} \times T_{on}I_{L-avg} = \frac{Vpsin\varpi t}{2*L_m} \times {T_{on}}^2 \times \frac{1}{T_{OSC}}$$

The peak current  $I_{L\_peak}$  and average current  $I_{L\_avg}$  of transformer will be shaped as AC input sinusoid too because  $(T_{ON})$  and  $f_{OSC}$  both are constant and then high power factor can be achieved.

## Real Current Regulator without Secondary Feedback

The proprietary real current control method allows the UTC **UCL5811** to accurately control the secondary side LED current from the primary side informatio. The output LED mean current can be calculated approximately as:

$$I_{OUT} = \frac{1}{2} \times \frac{1}{5*R_{CS}} \times V_{REF} \times \frac{N_p}{N_s}$$

Where  $I_{OUT}$  is the secondary output current of LED,  $V_{REF}$  is the inner reference voltage.  $N_P$  is number of turns of primary winding and  $N_S$  is number of turns of the secondary winding.

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