



## USR3652S

Preliminary

CMOS IC

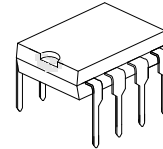
### HIGH PRECISION CC/CV PRIMARY-SIDE PWM POWER SWITCH

#### ■ DESCRIPTION

The UTC **USR3652S** is a primary control switch mode charger and adapter applications. The controlled variable is transferred by an auxiliary winding from the secondary to the primary side. The device integrates PWM controller to enhance the performance of discontinuous conduction mode (DCM) flyback converters.

The UTC **USR3652S** operates in primary-side sensing and regulation. Opto-coupler and TL431 could be eliminated. It also provides off-time modulation to linearly decrease PWM frequency under light-load conditions so that low standby power can be achieved.

The UTC **USR3652S** achieves high precision CV/CC regulation and high power efficiency. It offers comprehensive protection coverage with auto-recovery features including Cycle-by-cycle current limiting,  $V_{DD}$  over voltage protection,  $V_{DD}$  clamp, OTP, leading edge blanking,  $V_{DD}$  under voltage lockout, etc.



DIP-8

#### ■ FEATURES

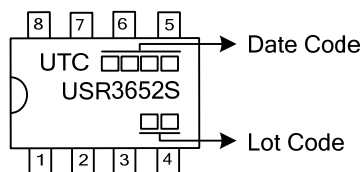
- \* High Voltage startup circuit
- \* Built-in 650V Power MOSFET
- \*  $\pm 5\%$  constant voltage regulation at universal AC input
- \* High precision constant current regulation at universal AC input
- \* Primary side control without TL431 and opto-coupler
- \* Programmable CV and CC regulation
- \* Programmable cable compensation in CV mode
- \* Flyback topology in DCM operation
- \* Frequency hopping to reduce system EMI
- \* Built-in leading edge blanking
- \* Built-in primary winding inductance compensation
- \* Power on soft start
- \* Gate Output Maximum Voltage Clamped at 15V
- \* Cycle-by-cycle current limiting
- \* Over voltage protection
- \*  $V_{DD}$  under-voltage lockout

#### ■ ORDERING INFORMATION

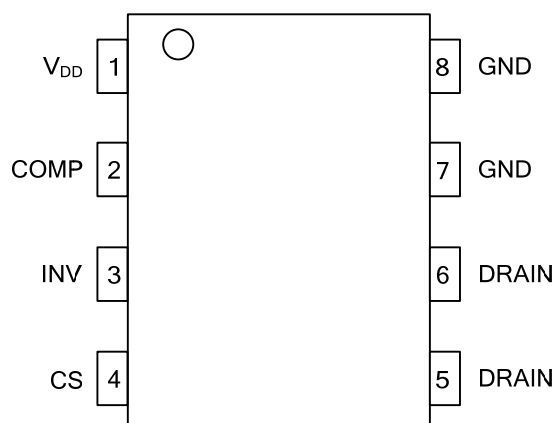
Ordering Number		Package	Packing
Lead Free	Halogen Free		
USR3652SL-D08-T	USR3652SG-D08-T	DIP-8	Tube

USR3652SG-D08-T	
(1)Packing Type	(1) T: Tube
(2)Package Type	(2) D08: DIP-8
(3)Green Package	(3) G: Halogen Free and Lead Free, L: Lead Free

### MARKING



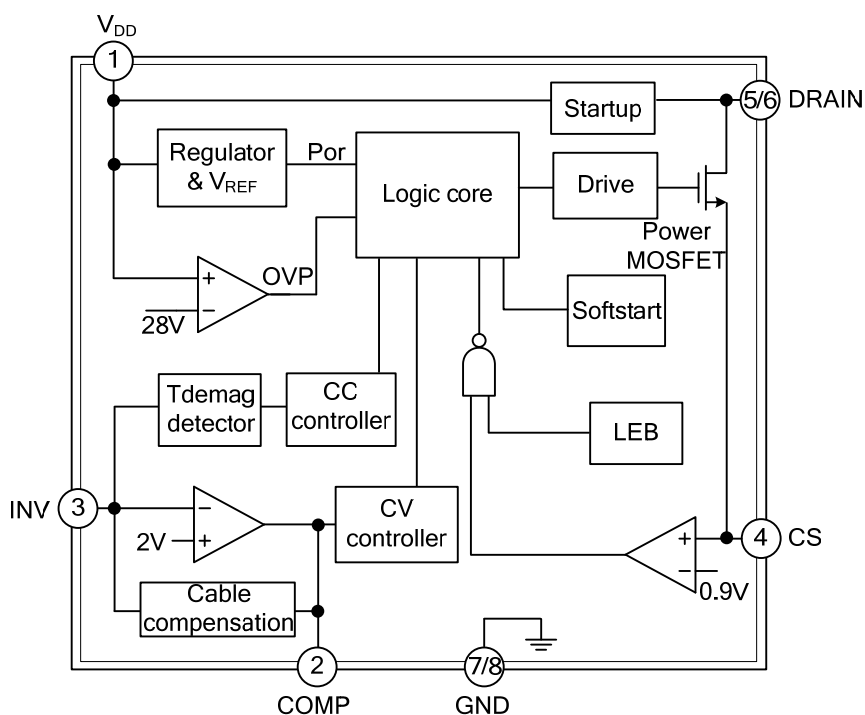
### PIN CONFIGURATION



### PIN DESCRIPTION

PIN NO.	PIN NAME	DESCRIPTION
1	V <sub>DD</sub>	Power supply
2	COMP	Loop compensation
3	INV	The voltage feedback from the auxiliary winding
4	CS	Current sense input
5/6	DRAIN	HV Power MOSFET Drain Pin. The Driain pin is connected to the primary lead of the transformer.
7/8	GND	Ground

### ■ BLOCK DIAGRAM



### ■ ABSOLUTE MAXIMUM RATING

PARAMETER	SYMBOL	RATINGS	UNIT
Drain Voltage (Off State)	$B_{VDSS}$	-0.3~ $B_{VDSS}$	V
$V_{DD}$ Zener Clamp Continuous Current	$I_{DD}$	10	mA
COMP Voltage	$V_{COMP}$	-0.3~7	V
CS Input Voltage	$V_{CS}$	-0.3~7	V
INV Input Voltage	$V_{INV}$	-0.3~7	V
Operating Junction Temperature	$T_J$	-20~150	°C
Storage Temperature	$T_{STG}$	-55~150	°C

Note: Absolute maximum ratings are those values beyond which the device could be permanently damaged.  
Absolute maximum ratings are stress ratings only and functional device operation is not implied.

### ■ THERMAL CHARACTERISTICS

PARAMETER	SYMBOL	RATINGS	UNIT
Junction to Ambient	$\theta_{JA}$	75	°C/W

### ■ OUTPUT POWER TABLE

PARAMETER		RATINGS	UNIT
Adapter (Note 1)	230VAC±15%	14	W
	85-264VAC	12	W

■ ELECTRICAL CHARACTERISTICS (T<sub>A</sub>=25°C, V<sub>DD</sub>=15V, unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>STARTUP SECTION</b>						
Startup Voltage	V <sub>st</sub>	V <sub>DD</sub> <V <sub>TH(ON)</sub>	25			V
Startup Current	I <sub>st</sub>	V <sub>DD</sub> <V <sub>TH(ON)</sub>		700		μA
<b>SUPPLY SECTION</b>						
Turn-on Threshold Voltage	V <sub>TH(ON)</sub>		13.5	14.5	16	V
Turn-off Threshold Voltage	V <sub>TH(OFF)</sub>		8.2	9	10.5	V
V <sub>DD</sub> OVP Protection	V <sub>DD_OVP</sub>			28		V
Operating Current	I <sub>DD-OP</sub>			1.6	2	mA
Green-Mode Operating Current	I <sub>DD-GREEN</sub>			1.3		mA
<b>OSCILLATOR SECTION</b>						
Maximum Frequency	f <sub>OSC</sub>		55	65	75	kHz
Minimum Frequency at CCM	f <sub>OSC-CM-MIN</sub>			17		kHz
Minimum Frequency at No-load	f <sub>OSC-N-MIN</sub>			370		Hz
Frequency Hopping Range	Δf/freq			±5.0		%
Frequency Hopping Period	t <sub>FHR</sub>			3		ms
<b>INV SECTION</b>						
Reference Voltage for EA	V <sub>VR</sub>		1.97	2.00	2.03	V
Reference Voltage for CCM	V <sub>VR_CCM</sub>			0.5		V
Max. Cable Compensation Current	I <sub>CABLE_MAX</sub>			37		μA
<b>CURRENT SENSE SECTION</b>						
Current Sense Threshold Voltage	V <sub>CS</sub>		880	910	940	mV
OCP Propagation Delay	t <sub>D_OC</sub>			90	200	nS
Leading Edge Blank Time	t <sub>LEB</sub>			900		nS
Soft Start Time	t <sub>SS</sub>			10		mS
<b>POWER MOSFET SECTION</b>						
MOSFET DS Breakdown Voltage	B <sub>VDSS</sub>		650			V
On Resistance	R <sub>DS(ON)</sub>			5		Ω
<b>OTP SECTION</b>						
Threshold Temperature of OTP (Note 2)	T <sub>OTP</sub>			+140		°C

Notes: 1. Maximum practical continuous power in an Adapter design with sufficient drain pattern as a heat sink, at 50°C ambient.

2. When the over-temperature protection is activated, the power system enters auto restart mode and output is disabled.

## ■ OPERATION DESCRIPTION

The UTC **USR3652S** is a primary control switch mode charger and adapter applications. It operates in primary-side sensing and regulation. Opto-coupler and TL431 could be eliminated. Proprietary built-in CV and CC control can achieve high precision CV/CC performance. The device integrates PWM controller to enhance the performance of discontinuous conduction mode (DCM) flyback converters.

### Startup control

The  $V_{DD}$  pin of UTC **USR3652S** is connected to the line input through a resistor. A large value startup resistor can be used to minimize the power loss in application because the start current of UTC **USR3652S** is very low. When the  $V_{DD}$  voltage reaches  $V_{TH(ON)}$ , the internal startup circuit is disabled and the IC turns on.

### Operating Current

The Operating current of UTC **USR3652S** is as low as 1.6mA. Good efficiency and very low standby power can be achieved.

### Constant Voltage Operation

The output voltage is defined by the transmission ratio between the secondary and auxiliary winding. The UTC **USR3652S** captures the auxiliary winding feedback voltage at INV pin and operates in constant-voltage (CV) mode to regulate the output voltage. The auxiliary voltage reflects the output voltage is given by:

$$V_{AUX} = \frac{N_A}{N_S} \times (V_O + \Delta V) \quad (1)$$

Where  $\Delta V$  indicates the drop voltage of the output diode.

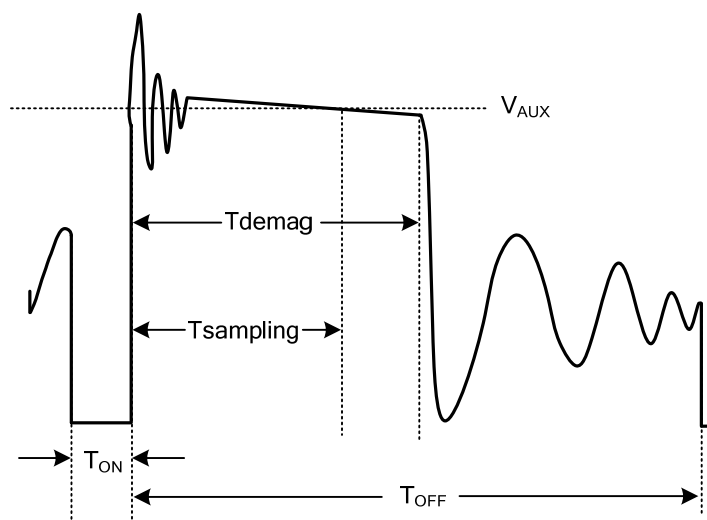


Figure 2. Auxiliary voltage waveform

Via a resistor divider connected between the auxiliary winding and INV, the  $V_{AUX}$  is sampled at the  $T_{sampling}$  end and it is hold until the next sampling. The sampled voltage is compared with 2.0V reference voltage and the error is amplified. The error amplifier output reflects the load condition and controls the  $T_{OFF}$  time and the  $I_{pk}$  to regulate the output voltage, thus constant output voltage can be achieved.

### Constant Current Operation

When the sampled voltage is below 2.0V reference voltage and the error amplifier output reaches its maximum, thus UTC **USR3652S** operates in constant-current (CC) mode. The CC point and maximum output power can be externally adjusted by external current sense resistor  $R_{cs}$ . The larger  $R_{cs}$ , the smaller CC point is, and the smaller output power becomes.

■ OPERATION DESCRIPTION (Cont.)

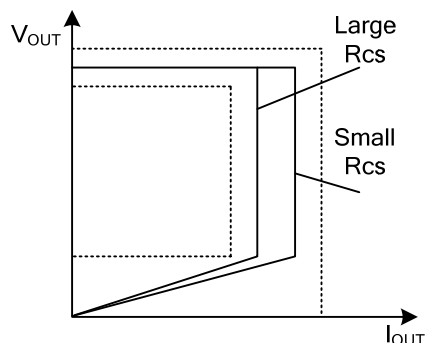


Figure 3. Adjustable output power by changing  $R_{CS}$

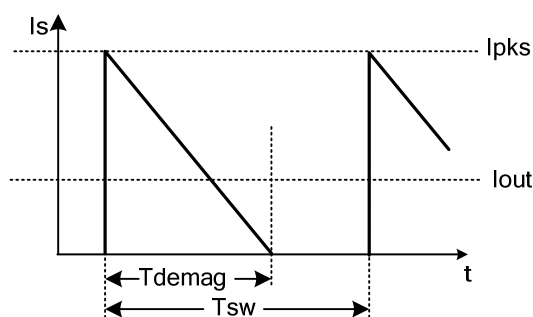


Figure 4. Secondary current waveform

In CC operation, the CC loop control function of UTC **USR3652S** will keep a fixed proportion between secondary inductance de-magnetization time ( $T_{demag}$ ) and switching cycle time ( $T_{sw}$ ). The fixed proportion is

$$\frac{T_{demag}}{T_{sw}} = \frac{1}{2} \quad (2)$$

Thus the output current is given by:

$$I_{OUT} = \frac{1}{2} \times \frac{N_P}{N_S} \times I_{PK} \times \frac{T_{demag}}{T_{sw}} = \frac{1}{4} \times \frac{N_P}{N_S} \times I_{PK} \quad (3)$$

### Programmable Cable Drop Compensation

UTC **USR3652S** has a built-in cable voltage drop compensation to achieve good load regulation. An offset voltage is generated at INV pin by an internal current flowing into the resistor divider. The current is inversely proportional to the voltage of COMP pin. As a result, it is inversely proportional to the output load current. The voltage drop across the cable is compensated by this offset voltage at INV pin. It can also be programmed by adjusting the resistance of the divider to compensate the drop for various cable lines used.

### Current Sensing and Leading Edge Blanking

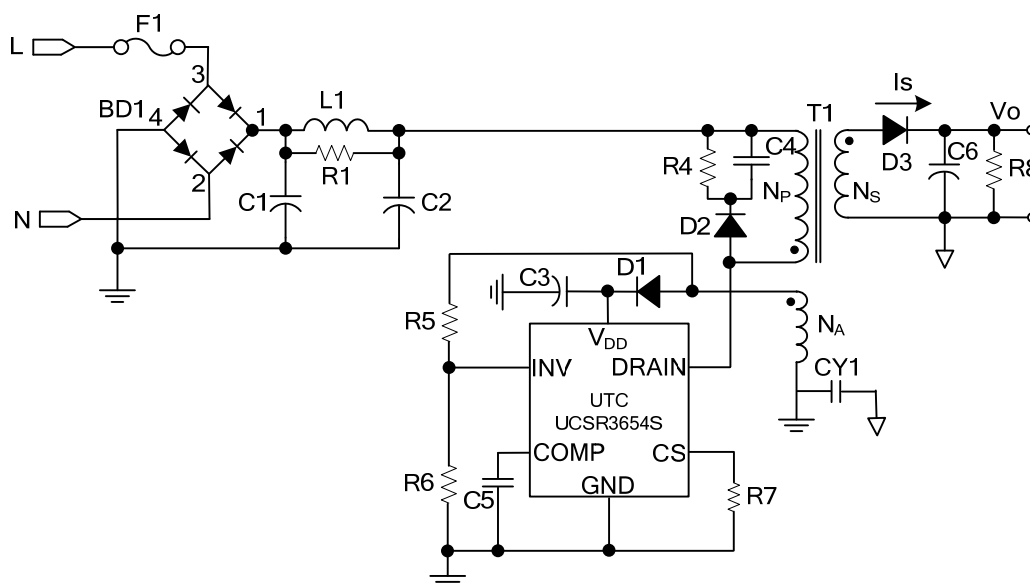
Cycle-by-cycle current limiting is offered in UTC **USR3652S**. The switch current is detected by a sense resistor into the CS pin. When the power switch is turned on, a turn-on spike will occur on this resistor. A 900ns leading-edge blanking is built in to avoid false-termination of the switching pulse so that the external RC filtering is no longer needed.

### Protection Control

Good power supply system reliability is achieved with its comprehensive protection features including  $V_{DD}$  over-voltage protection,  $V_{DD}$  Clamp, GATE Clamp, Power on soft start, Cycle-by-cycle current limiting, short circuit protection, leading edge blanking, OTP and UVLO, etc.

$V_{DD}$  is supplied by transformer auxiliary winding output. The output of UTC **USR3652S** is shutdown when  $V_{DD}$  drops below  $V_{TH(OFF)}$  and the power converter enters power on start-up sequence thereafter.

### ■ TYPICAL APPLICATION CIRCUIT (5V, 2.4A)



### BOM

Reference	Component	Reference	Component
BD1	BD 2A/650V	R1	R 2.4KΩ 1206 ±5%
L1	1m H 6x8mm	R4	R 200KΩ 1206 ±5%
C1	EC 10u F 400V 105°C	R5	R 18KΩ 0805 ±1%
C2	EC 22u F 400V 105°C	R6	R 3.3KΩ 0805 ±1%
C3	EC 4.7u F 50V 105°C	R7	R 1.2Ω 1206 ±5%
C4	CC 1n F 1000V 1206	R8	R 820Ω 1206 ±5%
C5	CC 0.1u F 50V 0805	T1	EE-19 or RM6
C6	EC 2200u F 10V 105°C	F1	FUSE 1A 250VAC
D1,D2	Diode UTC 1N4007G	CY1	Y/C 1000p F 250V
D3	Diode UTC MGBR40V45C		

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